



Voltage Regulator Module (VRM) and Enterprise Voltage Regulator-Down (EVRD) 10.0

Design Guidelines

July 2005



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Revision History

Date	Revision	Description
June 2004	001	Initial release.
June 2005	002	Added references to Low Voltage Intel® Xeon™ processor with 800 MHz system bus



1 Applications

This document defines DC-to-DC converters designed to help meet the power requirements of the Intel® Xeon™ processor with 800 MHz system bus and Low Voltage Intel® Xeon™ processor with 800 MHz system bus. The intent of this document is to define the electrical, thermal, and mechanical specifications for VRM 10.0.

VRM – The Voltage Regulator Module (VRM) designation in this document refers to a voltage regulator that is plugged into a baseboard where the baseboard is designed to support more than one processor. VRM output requirements in this document are intended to match the needs of a set of microprocessors.

EVRD – The Enterprise Voltage Regulator-Down (EVRD) designation in this document refers to a voltage regulator that is embedded in a baseboard where the baseboard is designed to support more than one processor. EVRD output requirements in this document are intended to match the needs of a set of microprocessors. Each implementation of a specific baseboard must meet the specifications of all processors supported by the baseboard.

The specifications in the respective processors' datasheet always take precedence over the data provided in this document.

VRM/EVRD 10.0 incorporates functional changes from prior VRD and VRM guidelines:

- Vcc and Vtt supplies are separate (the Vtt specifications are included in the respective processors' datasheet).
- Addition of a continuous load current (ICC (TDC)) to enable cost effective VRM/EVRD designs (Section 2.1).
- Addition of dynamic VID to change the output voltage during normal operation in response to an input from the processor (Section 2.7).
- Simplified definition of power-good as a power-up indication (Section 6.1).
- Extended use of the VRM/EVRD disable function to turn off the output in response to an input from the processor (Section 3.1).
- Specifications added to monitor and react to excessive temperature in the voltage regulator (VR) (Section 6.2)

1.1 Terminology

A '#' symbol after a signal name refers to an active low signal, indicating a signal is in the asserted state when driven to a low level. For example, when VRM_pres# is low, it indicates that a VRM is present in the connector.

Table 5 specifies the voltage level corresponding to the state of VID[5:0]. A '1' in this table refers to a high voltage level and a '0' refers to a low voltage level.



2 Output Voltage Requirements

2.1 Voltage and Current

REQUIRED

A 6-bit VID code provided by the processor to the VRM/EVRD determines a reference output voltage, as described in Section 3.2. Sections 2.2 and 2.3 specify deviations from the VID reference voltage.

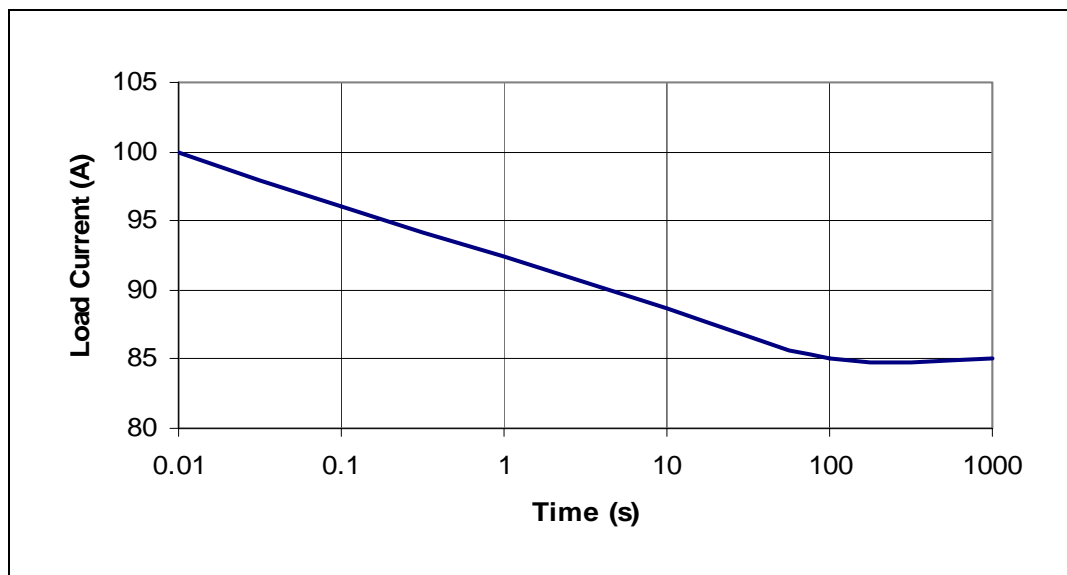
The load lines in Section 2.2 show the relationship between V_{cc} and I_{cc} for the processor.

For the Intel® Xeon™ processor with 800MHz system bus, the VRM/EVRD will be required to support the following:

- A continuous load current ($I_{cc}(TDC)$) of 85A
- A maximum load current ($I_{cc}(Max)$) of 100A
- A maximum load current step ($I_{cc}(Step)$), within a 1 μs period, of 70A
- A maximum current slew rate at the pins of the processor of 560A/ μs

Figure 1 displays the load current requirements over time.

Figure 1. VRM/EVRD 10.0 Load Current vs. Time for Intel® Xeon™ Processor with 800 MHz System Bus

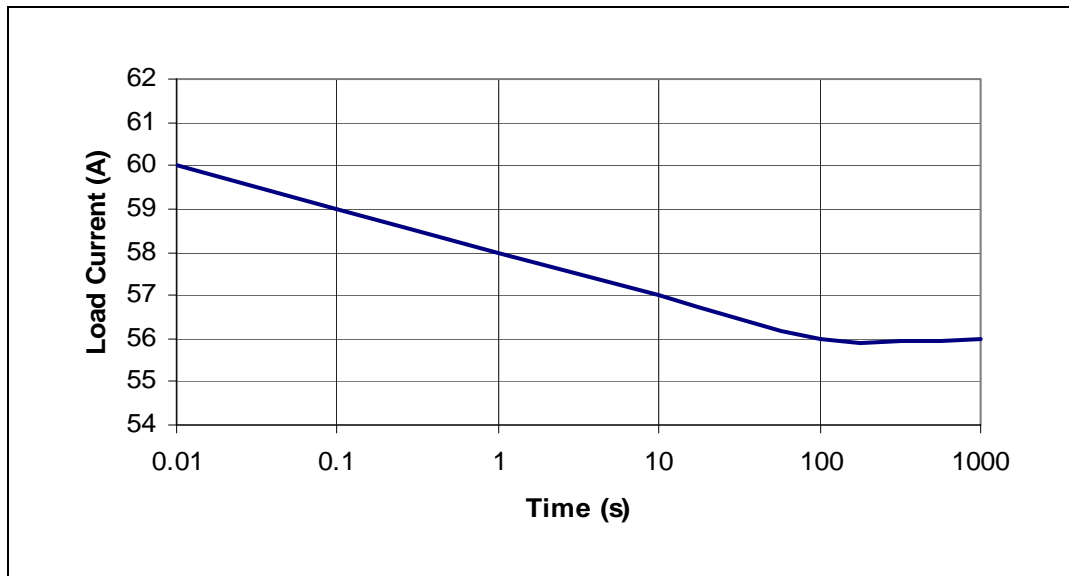


For the Low Voltage Intel® Xeon™ processor with 800MHz system bus, the VRM/EVRD will be required to support the following:

- A continuous load current ($I_{cc}(TDC)$) of 56A
- A maximum load current ($I_{cc}(Max)$) of 60A
- A maximum load current step ($I_{cc}(Step)$), within a 1 μs period, of 38.5A
- A maximum current slew rate at the pins of the processor of 308A/ μs

Figure 2 displays the load current requirements over time.

Figure 2 VRM/EVRD 10.0 Load Current vs. Time for Low Voltage Intel® Xeon™ Processor with 800 MHz System Bus



The continuous load current can also be referred to as the thermal design current (TDC). TDC is the sustained (DC equivalent) current that the processor is capable of drawing indefinitely and defines the current to use for the voltage regulator temperature assessment. At TDC, switching FETs reach maximum temperature and may heat the baseboard layers and neighboring components above valid thermal limits. Actual component and baseboard temperatures are established by the envelope of the system operating conditions. This includes voltage regulator layout, processor fan selection, ambient temperature, chassis configuration, etc. To avoid heat related failures, baseboards should be validated for thermal compliance under the envelope of system operating conditions.

The maximum load current represents the maximum peak current that the processor is capable of drawing. It is the maximum current the VRM/EVRD must be electrically designed to support without tripping any protection circuitry.

Table 1 lists the I_{cc} guidelines for the flexible motherboard (FMB) frequency of the processor. For designers who choose to design their VR thermal solution to the TDC, it is recommended that voltage regulator thermal protection also be implemented (see Section 6.2).



Table 1. Icc Guidelines

	Icc(TDC) (A)	Icc(Max) (A)	Icc(Step) (A)	Slew Rate (A/μs)
Intel® Xeon™ Processor with 800 MHz System Bus	85	100	70	560
Low Voltage Intel® Xeon™ processor with 800 MHz system bus	56	60	38.5	308

NOTES:

1. See the *Intel® Xeon™ Processor with 800 MHz System Bus Datasheet* and *Low Voltage Intel® Xeon™ processor with 800 MHz System Bus Datasheet* for the latest specifications.
2. This table represents the expected operating limitations for VRM/EVRD 10.0. Future Intel® Xeon™ processors with 800 MHz system bus will require a later version of the VRM/EVRD to reach end-of-life capabilities.

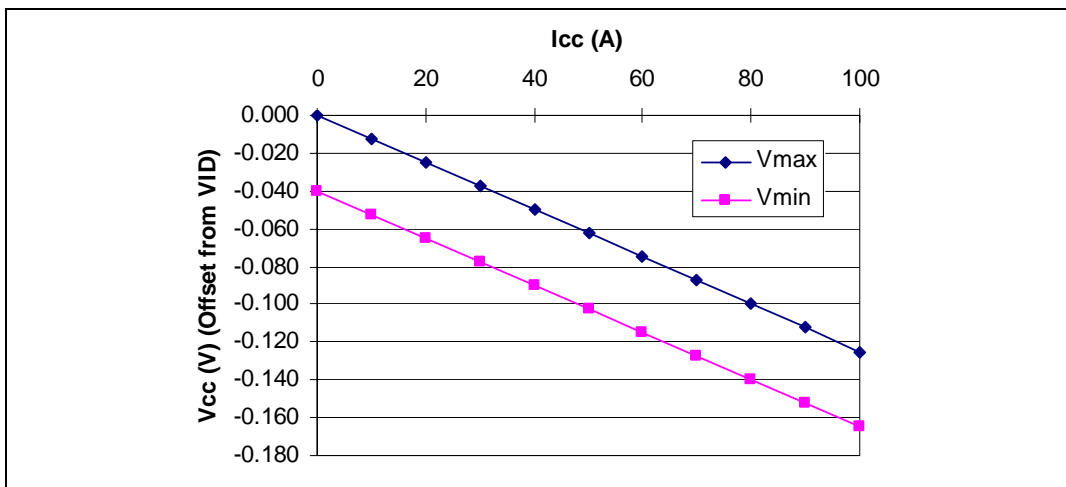
2.2 Load Line Definitions REQUIRED

The following load lines contain static and transient voltage data as well as maximum and minimum voltage levels. Measurement of the load line is to be done at the remote sense point. It is required that the remote sense point for the voltage regulator be connected to the processor VCCSENSE and VSSSENSE pins.

The upper and lower load lines represent the allowable range of voltages that must be presented to the processor. The voltage must never exceed these boundaries for proper operation of the processor.

Figure 3 shows load line voltage offsets and current levels based on the VID specifications.

Figure 3. VRM/EVRD 10.0 Die Load Line



The following equations for the load lines are valid for the range of load current from 0 to 100A.

- V_{MAX} load line: $V_{cc} = VID - (1.25 \text{ m}\Omega \cdot I_{cc})$
- V_{MIN} load line: $V_{cc} = VID - 0.040V - (1.25 \text{ m}\Omega \cdot I_{cc})$

2.3 Voltage Tolerance

REQUIRED

The voltage ranges shown in Section 2.2 include the following tolerances:

- Initial DC output voltage set-point error
- Output ripple and noise
- No-load offset centering error
- Current sensing and droop errors
- Component aging effects
- Full ambient temperature range and warm up
- Dynamic output changes from minimum-to-maximum or maximum-to-minimum load, as measured over a 20 MHz bandwidth
- Variations of the input voltage

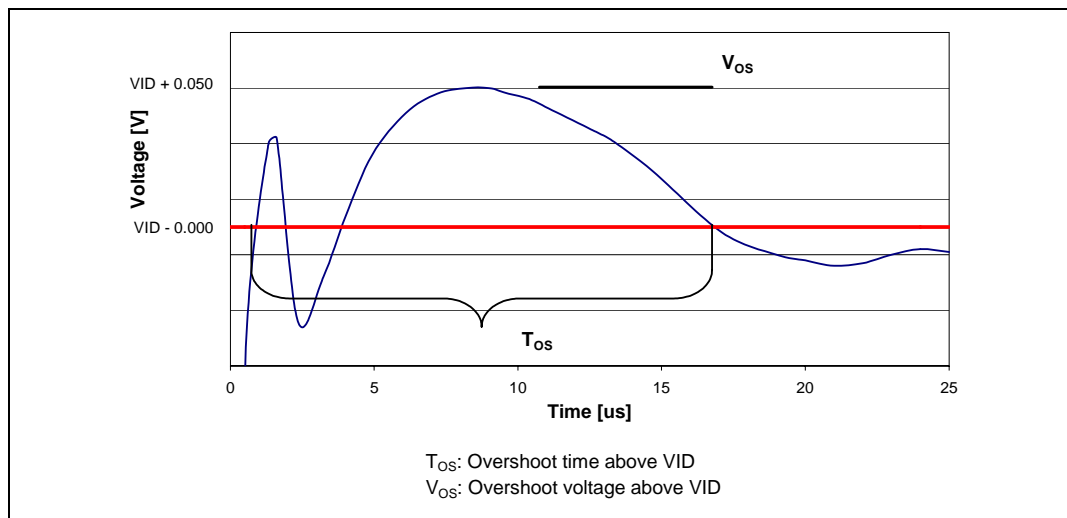
2.4 Processor Vcc Overshoot

REQUIRED

The Intel® Xeon™ processor with 800 MHz system bus processor and Low Voltage Intel® Xeon™ processor with 800 MHz system bus can tolerate short transient overshoot events where Vcc exceeds the VID voltage when transitioning from a high-to-low current load condition. This overshoot cannot exceed $VID + V_{OS_MAX}$. The overshoot duration, which is the time that the overshoot can remain above VID, cannot exceed T_{OS_MAX} . These specifications apply to the processor socket voltage as measured across the remote sense points and should be taken with a 20 MHz bandwidth limited oscilloscope.

- V_{OS_MAX} = Maximum overshoot voltage above VID = 50 mV
- T_{OS_MAX} = Maximum overshoot time duration above VID = 25 μ s

Figure 4. Vcc Overshoot Example Waveform





2.5 Stability

REQUIRED

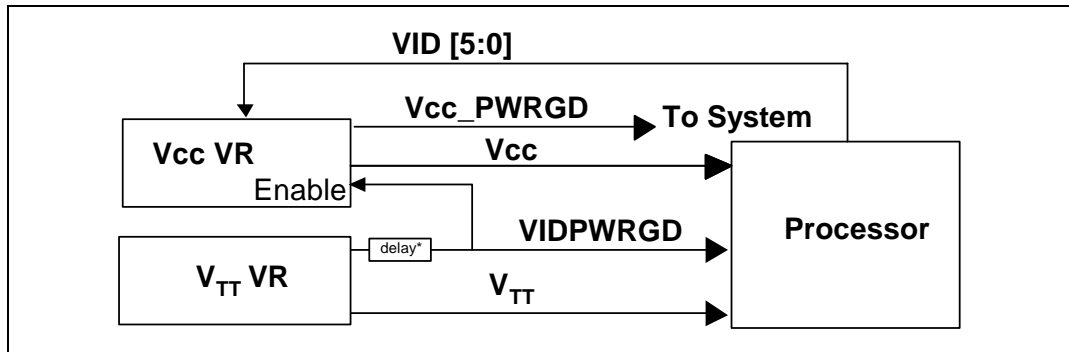
The VRM/EVRD needs to be unconditionally stable under all specified output voltage ranges and current transients of any duty cycle and up to repetition rates of 1 MHz. The VRM/EVRD should be stable under a no load condition.

2.6 Processor Power Sequencing

REQUIRED

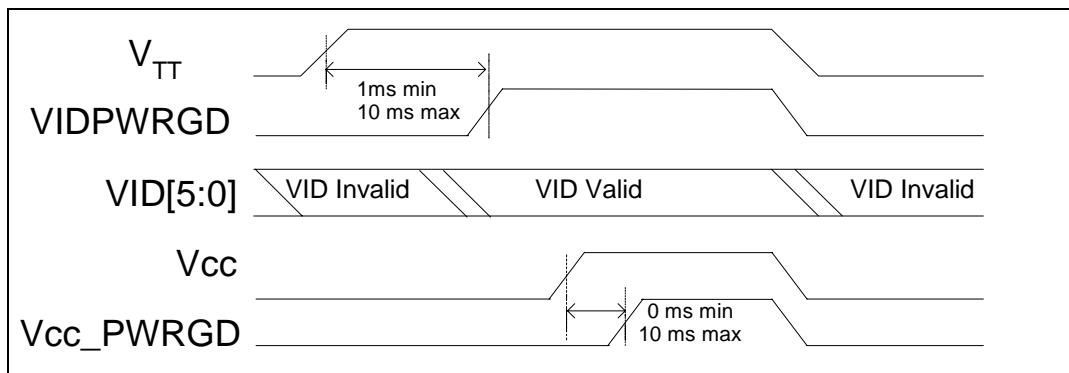
The VRM/EVRD must support platforms with defined power-up sequences. Figure 5 shows a block diagram of a power sequencing implementation, and Figure 6 shows a timing diagram of the power sequencing requirements.

Figure 5. Power-On Sequence Block Diagram



NOTES: This delay could be built into the V_{TT} VR.

Figure 6. Power-On Sequence Timing Diagram



NOTES:

1. V_{TT} comes up at the application of system power to the V_{TT} VR. V_{TT} is used to supply power to the VID logic of the processor.
2. V_{TT} VR generates VIDPWRGD, to latch the processor's VID outputs and enable the Vcc VR, after the V_{TT} supply reaches 90% of the final value.
3. Vcc_PWRGD is generated by the Vcc VR and may be used elsewhere in the system.
4. VIDPWRGD must deactivate and Vcc must be disabled immediately when Vtt becomes invalid.
5. Vcc_PWRGD should assert between 0 and 10 ms after Vcc reaches 90% of the final value.
6. See the *Intel® Xeon™ Processor with 800 MHz System Bus Datasheet* and *Low Voltage Intel® Xeon™ processor with 800 MHz System Bus Datasheet* for the latest timing requirements.

2.7 Dynamic Voltage Identification (VID) **REQUIRED**

VRM/EVRD 10.0 supports dynamic VID across the entire VID table, which requires the ability to reduce the load line voltage shown in Figure 3 by 450 mV. The VRM/EVRD must be capable of accepting voltage level changes of 12.5 mV steps every 5 μ s, up to 36 steps (450 mV) in 180 μ s. The low voltage state will be maintained for at least 50 μ s. The worst case settling time for the six VID lines, including line-to-line skew, is 400 ns. The VID inputs should contain circuitry to prevent false tripping or latching of VID codes during the settling time.

During a transition the output voltage must be between the maximum voltage of the high range (“A” in Figure 7) and the minimum voltage of the low range (“B”). The VRM/EVRD must respond to a transition from VID-low to VID-high by regulating its V_{cc} output to the range defined by the new, final VID code within 50 μ s of the final step. The time to move the output voltage from VID-high to VID-low will depend on the PWM controller design, the amount of system decoupling capacitance, and the processor load.

Figure 7 shows operating states as a representative processor changes levels. The diagram assumes steady state, maximum current during the transition for ease of illustration; actual processor behavior allows for any di/dt event during the transitions, depending on the code it is executing at that time. In the example, the processor begins in a high load condition. In transitions 1 \rightarrow 2 and 2 \rightarrow 3, the processor prepares to switch to the low voltage range with a transition to a low load condition, followed by an increased activity level. Transition 3 \rightarrow 4 is a simplification of the multiple steps from the high voltage load line to the low voltage load line. Transition 4 \rightarrow 5 is an example of a response to a load change during normal operation in the lower range.

Figure 7. Processor Transition States

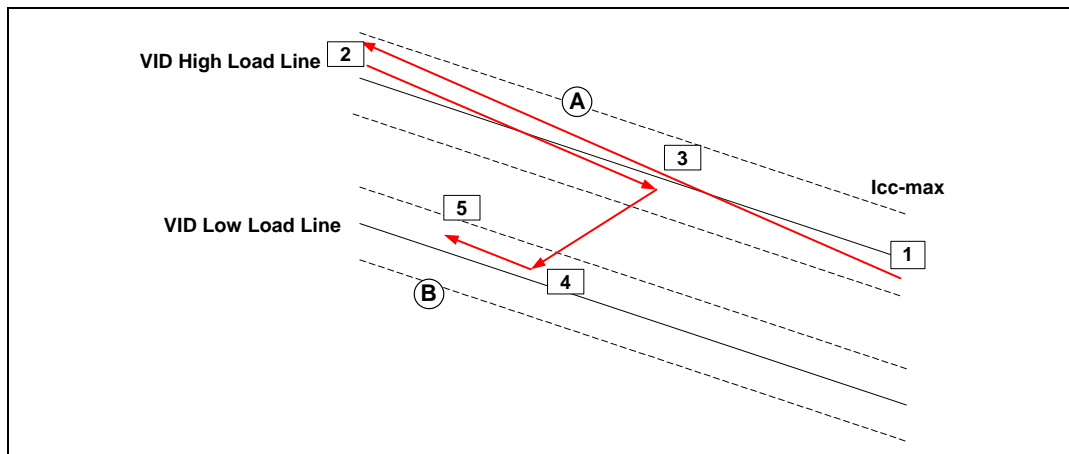
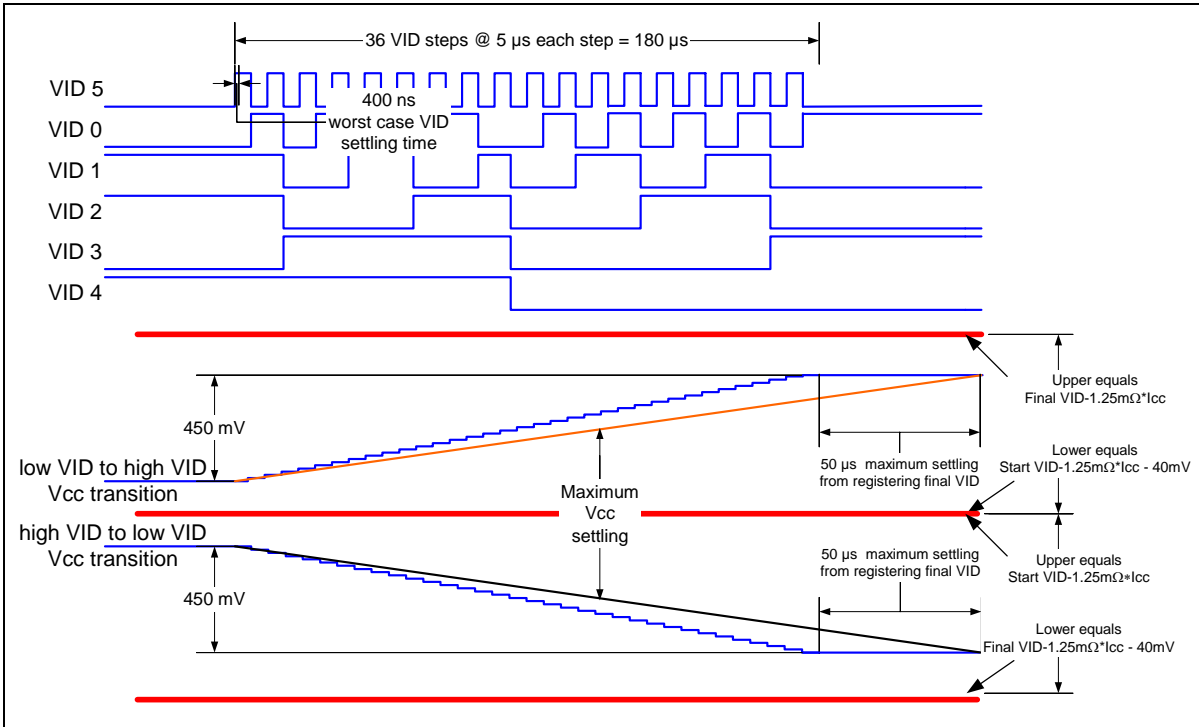


Figure 8 provides an illustration of dynamic VID. The diagram in Figure 8 assumes steady state, constant current during the dynamic VID transition for ease of illustration; actual processor behavior allows for any dI_{cc}/dt during the transitions, depending on the code it is executing at that time. Note that during dynamic VID, the processor will not output VID codes that would disable the voltage regulator output voltage.

The processor load may not be sufficient to absorb all of the energy from the output capacitors on the baseboard when VID changes to a lower output voltage. The VRM/EVRD design should ensure that any energy transfer from the capacitors does not impair the operation of the VRM/EVRD, the AC-DC supply, or any other parts of the system.



Figure 8. Example Processor VID Transition States



NOTE: For ease of illustration, the diagram assumes steady state, constant current during the dynamic VID Transition. Actual processor behavior allows for any $dlcc/dt$ event during the transitions, depending on the code it is executing at that time.

2.8 Overshoot at Turn-On or Turn-Off REQUIRED

The VRM/EVRD output voltage should remain within the load-line regulation band for the VID setting while the VRM/EVRD is turning on or turning off, with no over- or undershoot out of regulation. No negative voltage below -100 mV may be present at the VRM/EVRD output during turn-on or turn-off.

2.9 Output Filter Capacitance REQUIRED

The output filter capacitance for VRM based designs will be located on the baseboard. The system design must ensure that the output voltage of the VR conforms to the load line of Figure 3 with the baseboard and processor loads.

Figure 9 is a representative example of a baseboard decoupling solution and a processor load. The values shown are for a four-phase switching voltage regulator design. The parasitic baseboard values are extracted from a design using a four-layer baseboard with 2 ounces total of copper for Vcc and 2 ounces total of copper for ground. The amount of bulk decoupling required is dependent on the voltage regulator design.

This design incorporates at least ten 560 μF Aluminum-polymer bulk capacitors and forty 10 μF ceramic high frequency capacitors per processor (Table 2). Eight of the 10 μF capacitors should be placed in the cavity of the processor socket. The remaining 32 capacitors should be split evenly such that half are on one side of the processor socket and half are on the other side as close

to the processor socket as the keepout zones allow. If backside passive components are allowed in the design, it would be beneficial to place the 32 capacitors under the processor socket on the backside of the baseboard. Five of the 560 μF capacitors should be placed on one side of the processor socket and five on the other side as close to the processor socket as the keepout zones allow.

Note: The amount of bulk decoupling needed is dependent on the voltage regulator design. Some multiphase buck regulators may have a higher switching frequency that would require a different output decoupling solution to meet the processor load line requirements than that described in this document.

The impedance values labeled “Socket and Package Pins” are supplied by Intel Corporation and are beyond the control of the system designer. They represent the mPGA604 socket and the processor package.

For VRM applications, it is recommended that the system designer work with the VRM supplier to ensure proper implementation of the VRM converter.

Figure 9. Model of Processor Load

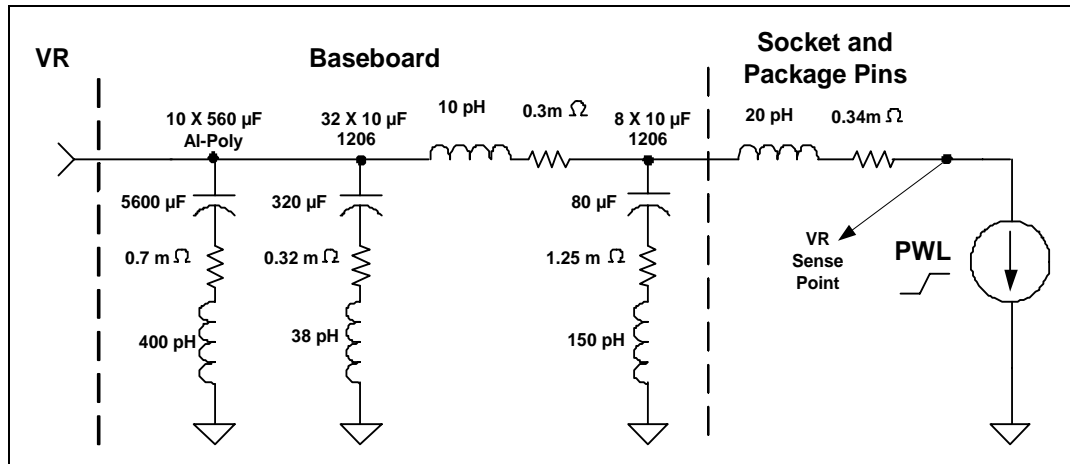


Table 2. Capacitor Recommendations

Qty	Value	Tolerance	Temp Coeff	ESR (m Ω)	ESL (nH)	Notes
10	560 μF Al-Polymer	$\pm 20\%$	N/A	7	4.0	
8	10 μF Ceramic	$\pm 20\%$	X6S	10	1.2	1
32	10 μF Ceramic	$\pm 20\%$	X5R or X6S	10	1.2	

NOTES:

1. This row corresponds to the capacitors in the processor socket cavity.

2.10 Shut-Down Response

REQUIRED

Once the VRM/EVRD is operating after power-up, if either the Output Enable signal is deasserted or VID[5:0] = X11111, the VRM/EVRD should turn off its output (the output should go to high impedance) within 500 ms.

3 Control Signals

3.1 Output Enable (OUTEN)

REQUIRED

The VRM/EVRD should accept an input signal to enable its output voltage. When disabled, the VRM/EVRD output voltage should go to a high impedance state and should not sink or source current. When OUTEN is pulled low during the shutdown process, the VRM/EVRD should not exceed the previous voltage level regardless of the VID setting during the shutdown process. Once the VRM/EVRD is operating after power-up, it should respond to a deasserted OUTEN within 500 ms. The circuitry driving OUTEN is an open-collector signal. It is **EXPECTED** that the pull-up resistor will be located on the baseboard and will not be integrated into the PWM controller chip.

Table 3. OUTEN Specifications

Symbol	Parameter	Min	Max	Units
V _{IH}	Input Voltage High	0.8	3.465	V
V _{IL}	Input Voltage Low	0.0	0.400	V

3.2 Voltage Identification (VID[5:0])

REQUIRED

VID [4:0] are compatible with Intel® Pentium® 4 and Intel® Xeon™ processors using 5-bit VID codes. VID [5:0] will be used on processors with 6-bit VID codes.

The VRM/EVRD must accept six lines to set the nominal voltage as defined in this section. When the VID[4:0] inputs are all high (in this case VID5 is a don't care), such as when no processor is installed, the VRM/EVRD should disable its output voltage. If this disable code appears during previously normal operation, the VRM/EVRD should turn off its output voltage within 500 ms. The circuitry driving VID[5:0] is an open-collector signal. It is **EXPECTED** that the pull-up resistors will be located on the baseboard and will not be integrated into the PWM controller chip. Other platform components may use VID inputs and may require tighter limits than specified in Table 4.

A normal no-processor VID [5:0] code for a V_{cc} regulator will be X11111, where X is defined as logic 1 or 0, disabling the VRM/EVRD.

Table 4. VID Specifications

Symbol	Parameter	Min	Max	Units	Notes
V _{IH}	Input High Voltage	0.8	3.465	V	1, 2
V _{IL}	Input Low Voltage	0	0.4	V	1, 2

NOTES:

1. Other platform components may use VID inputs and may require tighter limits.
2. These specifications are for the VRM/EVRD. They are not processor specifications.



Table 5. Voltage Identification (VID)

Processor Pins (0 = low, 1 = high)						Vout (V)	Processor Pins (0 = low, 1 = high)						Vout (V)
VID4	VID3	VID2	VID1	VID0	VID5		VID4	VID3	VID2	VID1	VID0	VID5	
0	1	0	1	0	0	0.8375	1	1	0	1	0	0	1.2125
0	1	0	0	1	1	0.8500	1	1	0	0	1	1	1.2250
0	1	0	0	1	0	0.8625	1	1	0	0	1	0	1.2375
0	1	0	0	0	1	0.8750	1	1	0	0	0	1	1.2500
0	1	0	0	0	0	0.8875	1	1	0	0	0	0	1.2625
0	0	1	1	1	1	0.9000	1	0	1	1	1	1	1.2750
0	0	1	1	1	0	0.9125	1	0	1	1	1	0	1.2875
0	0	1	1	0	1	0.9250	1	0	1	1	0	1	1.3000
0	0	1	1	0	0	0.9375	1	0	1	1	0	0	1.3125
0	0	1	0	1	1	0.9500	1	0	1	0	1	1	1.3250
0	0	1	0	1	0	0.9625	1	0	1	0	1	0	1.3375
0	0	1	0	0	1	0.9750	1	0	1	0	0	1	1.3500
0	0	1	0	0	0	0.9875	1	0	1	0	0	0	1.3625
0	0	0	1	1	1	1.0000	1	0	0	1	1	1	1.3750
0	0	0	1	1	0	1.0125	1	0	0	1	1	0	1.3875
0	0	0	1	0	1	1.0250	1	0	0	1	0	1	1.4000
0	0	0	1	0	0	1.0375	1	0	0	1	0	0	1.4125
0	0	0	0	1	1	1.0500	1	0	0	0	1	1	1.4250
0	0	0	0	1	0	1.0625	1	0	0	0	1	0	1.4375
0	0	0	0	0	1	1.0750	1	0	0	0	0	1	1.4500
0	0	0	0	0	0	1.0875	1	0	0	0	0	0	1.4625
1	1	1	1	1	1	OFF ¹	0	1	1	1	1	1	1.4750
1	1	1	1	1	0	OFF ¹	0	1	1	1	1	0	1.4875
1	1	1	1	0	1	1.1000	0	1	1	1	0	1	1.5000
1	1	1	1	0	0	1.1125	0	1	1	1	0	0	1.5125
1	1	1	0	1	1	1.1250	0	1	1	0	1	1	1.5250
1	1	1	0	1	0	1.1375	0	1	1	0	1	0	1.5375
1	1	1	0	0	1	1.1500	0	1	1	0	0	1	1.5500
1	1	1	0	0	0	1.1625	0	1	1	0	0	0	1.5625
1	1	0	1	1	1	1.1750	0	1	0	1	1	1	1.5750
1	1	0	1	1	0	1.1875	0	1	0	1	1	0	1.5875
1	1	0	1	0	1	1.2000	0	1	0	1	0	1	1.6000

NOTE:

3. Output disabled – the same as deasserting the Output Enable input (Section 3.1).



3.3 Differential Remote Sense (VO_SEN+/-) REQUIRED

The PWM controller will include differential sense inputs to compensate for an output voltage offset of <300 mV in the power distribution path. This common mode voltage is expected to occur due to transient currents and parasitic inductances and is not expected to be caused by parasitic resistances. The remote sense lines should draw no more than 10 mA, to minimize offset errors.

3.4 VRM Present (VRM_pres#) EXPECTED

The VRM should have the VRM_pres# signal. This signal is an output signal used to indicate to the system that a VRM is plugged into the socket. VRM_pres# is an open-collector/drain or equivalent signal. Table 6 shows the VRM_pres# pin specification. It is **EXPECTED** that the pull-up resistor will be located on the baseboard and will not be integrated into the VRM.

Table 6. VRM_pres# Specifications

Symbol	Parameter	Min	Max	Units
I _{OL}	Output Low Current	0	4	mA
V _{OH}	Output High Voltage	0.8	5.5	V
V _{OL}	Output Low Voltage	0	0.4	V



4 **Input Voltage and Current**

4.1 **Input Voltages**

EXPECTED

The power source for the VRM/EVRD is 12 V +5/-8%. This voltage is supplied by a separate power supply. For input voltages outside the normal operating range, the VRM/EVRD should either operate properly or shut down.

4.2 **Load Transient Effects on Input Current** **EXPECTED**

The design of the VRM/EVRD, including the input power delivery filter must ensure that the maximum slew rate of the input current does not exceed 0.5 A/ μ s, or as specified by the separate power supply.

Note: *In the case of a VRM design, the input power delivery filter may be located either on the VRM or on the baseboard. The decision for the placement of the filter will need to be coordinated between the baseboard and VRM designers.*

It is recommended that the bulk input decoupling be placed on the baseboard by the VRM input connector and high frequency decoupling on the VRM module. The expected baseboard decoupling should be between 1000 μ F to 2240 μ F depending on the VRM design and system power supply.



5 **Processor Voltage Output Protection**

These are features built into the VRM/EVRD to help prevent damage to itself, the processor, or other system components.

5.1 **Over-Voltage Protection (OVP) PROPOSED**

The OVP circuit monitors the processor core voltage (V_{cc}) for an over-voltage condition. If the output is more than 200 mV above the VID level, the VRM/EVRD shuts off the output.

5.2 **Over-Current Protection (OCP) PROPOSED**

The VRM/EVRD should be capable of withstanding a continuous, abnormally low resistance on the output without damage or over-stress to the unit. Output current under this condition will be limited to no more than 120% of the maximum rated output of the voltage regulator at thermal equilibrium.



6 Output Indicators

6.1 Vcc Power-Good (Vcc_PWRGD) PROPOSED

The VRM/EVRD may provide a power-good output signal, which remains in the low state until a maximum of 10 ms after the output voltage reaches the range specified in Section NOTES:22.2. The signal should then remain asserted as long as the VRM/EVRD output is operating within specification. It will be an open-collector/drain or equivalent signal. The pull-up resistor and voltage source will be located on the baseboard. If this signal is not implemented on the VRM, it should be left unconnected.

Table 7. Vcc_PWRGD Specifications

Symbol	Parameter	Min	Max	Units
I _{OL}	Output Low Current	0.0	4.0	mA
V _{OH}	Output High Voltage	0.8	5.5	V
V _{OL}	Output Low Voltage	0.0	0.4	V

6.2 Voltage Regulator Hot (VR_hot#) PROPOSED

The VR_hot# is an output signal that is asserted low when a thermal event is detected in the VRM/EVRD. Assertion of this signal will be used by the system to minimize damage to the VRM/EVRD due to the thermal conditions. It will be an open-collector/drain or equivalent signal. The pull-up resistor and voltage source will be located on the baseboard. A typical implementation would be a 50 ohm resistor pulled up to 1.2V.

Table 8. VR_hot# Specifications

Symbol	Parameter	Min	Max	Units
I _{OL}	Output Low Current	19.9	30.0	mA
V _{OH}	Output High Voltage	0.8	3.465	V
V _{OL}	Output Low Voltage	0.0	0.40	V

Each customer is responsible for identifying maximum temperature specifications for all components in the VRM/EVRD design and ensuring that these specifications are not violated while continuously drawing specified I_{cc} (TDC) levels. In the occurrence of a thermal event, a thermal sense circuit is to assert the processor signal FORCEPR# immediately prior to exceeding maximum VRM, baseboard, and/or component thermal ratings to prevent heat damage. The assertion may be made through direct connection to the FORCEPR# pin or through system management logic. Assertion of this signal will lower processor power consumption and reduce current draw through the voltage regulator, resulting in lower component temperatures. Sustained assertion of the FORCEPR# pin may cause noticeable platform performance degradation and must never occur when drawing less than specified thermal design current.

It is recommended that hysteresis be designed into the thermal sense circuit to prevent a scenario in which the VR_hot# signal is rapidly being asserted and deasserted.

6.3 Load Indicator Output (Load_current) PROPOSED

The VRM/EVRD may have an output with a voltage level that varies linearly with the VRM/EVRD output current. The PWM controller supplier may specify a voltage-current relationship consistent with the controller's current sensing method. Baseboards may route this output to a test point for system validation.

6.4 VRM Identification (VRM_ID0, VRM_ID1) PROPOSED

The VRM identification signals must be connected to ground on the VRM10.0 module. These signals can be used in systems that are designed to accept both VRM10.0 and VRM10.1 modules. The intent is to provide the system designer with a method to ensure that the capabilities of the installed VRM match the power requirements of the processor.

These signals correspond to the Load Line Select (LL0, LL1) signals on the VRM10.1 module. For processors that require VRM 10.1, one or both of these Load Line Select signals will be tied to a pull-up resistor on the baseboard. If a VRM10.1 module is inserted into the connector, the signals will be pulled up as expected. However, if a VRM10.0 module is inserted into the connector, both signals will be low since they are tied to ground on the VRM10.0 module. The system can then take the appropriate action.

Some SKUs of the Intel® Xeon™ processor with 800 MHz system bus will work with either a VRM10.0 module or a VRM10.1 module. In this case the VRM identification signals are a don't care.

For more information on the Load Line Select signals, see the *Voltage Regulator Module (VRM) and Enterprise Voltage Regulator-Down (EVRD) 10.1 Design Guidelines*.

7 VRM – Mechanical Guidelines

7.1 VRM Connector

EXPECTED

The VRM interface with the baseboard is a 31-pin pair edge connector. The connector uses latches to hold the VRM in place. Table 9 shows the connector vendors and part numbers.

Table 9. VRM 10.0 Connector Vendor and Part Numbers

Vendor	Part Number	Notes
Tyco/Amp	1489930-2	1
Foxconn	2EV04607-NW	1

NOTES:

1. These vendors are listed by Intel as a convenience to Intel's general customer base, but Intel does not make any representations or warranties whatsoever regarding quality, reliability, functionality, or compatibility of these devices. This list and/or these devices may be subject to change without notice.

7.2 VRM Connector Keying

The connector contains a single notch between pins 10 and 11 (52 and 53 on opposite side) as shown in Figure 10. This configuration allows both VRM10.0 and VRM 10.1 modules to be plugged into the connector. The connector footprint key to the baseboard is at pin 57.

7.3 Pin Descriptions and Assignments

Table 10 shows the VRM10.0 connector pin definitions. The pins that are labeled Reserved should not be used if there are plans to use VRM10.1 modules with this connector. Those pins are used on the VRM10.1 module as an additional input voltage pair. Systems that also plan to use VRM10.1 modules should connect pin 4 to VIN+ and pin 59 to VIN-. See the *Voltage Regulator Module (VRM) and Enterprise Voltage Regulator-Down (EVRD) 10.1 Design Guidelines* for more information. Otherwise those pins can be classified as Unspecified. The pin assignments are shown in Table 11.

Table 10. VRM 10.0 Connector Pin Descriptions

Name	Type	Description
Load_current	Output	Analog signal representing the output load current.
OUTEN	Input	Output Enable.
Vcc_PWRGD	Output	Output signal indicating that the output voltage of the VRM is in the specified range.
VID[5:0]	Input	Voltage ID pins used to specify the VRM output voltage.
VIN+	Power	VRM Input Voltage.
VIN-	Ground	VRM Input Ground.
VO+	Power	VRM Output Voltage.
VO-	Ground	VRM Output Ground.
VO_SEN+ VO_SEN-	Input	Output voltage sense pins.
VR_hot#	Output	Indicates to the system that a thermal event has been detected in the VR
VRM_pres#	Output	Indicates to the system that a VRM is plugged into the socket.
VRM_ID0	Input	VRM Identification bit 0
VRM_ID1	Input	VRM Identification bit 1
Reserved	Reserved	Do not use if planning to use VRM10.1 modules with this connector.
Unspecified	Unspecified	Pins available for use by vendors or customers.

Table 11. VRM Pins

1	VIN+
2	VIN+
3	VIN+
4	Reserved
5	VR_hot#
6	VID4
7	VID3
8	VID1
9	Load_current
10	Vcc_PWRGD
11	VO_SEN-
12	VRM_ID0
13	VO-
14	VO+
15	VO-
16	VO+
17	VO-
18	VO+
19	VO-
20	VO+
21	VO-
22	VO+
23	VO-
24	VO+
25	VO-
26	VO+
27	VO-
28	VO+
29	VO-
30	VO+
31	VO-
62	VIN-
61	VIN-
60	VIN-
59	Reserved
58	VRM_pres#
57	Key
56	VID2
55	VID0
54	VID5
53	OUTEN
52	VO_SEN+
51	VRM_ID1
50	VO+
49	VO+
48	VO-
47	VO+
46	VO-
45	VO+
44	VO-
43	VO+
42	VO-
41	VO+
40	VO-
39	VO+
38	VO-
37	VO+
36	VO-
35	VO+
34	VO-
33	VO+
32	VO-

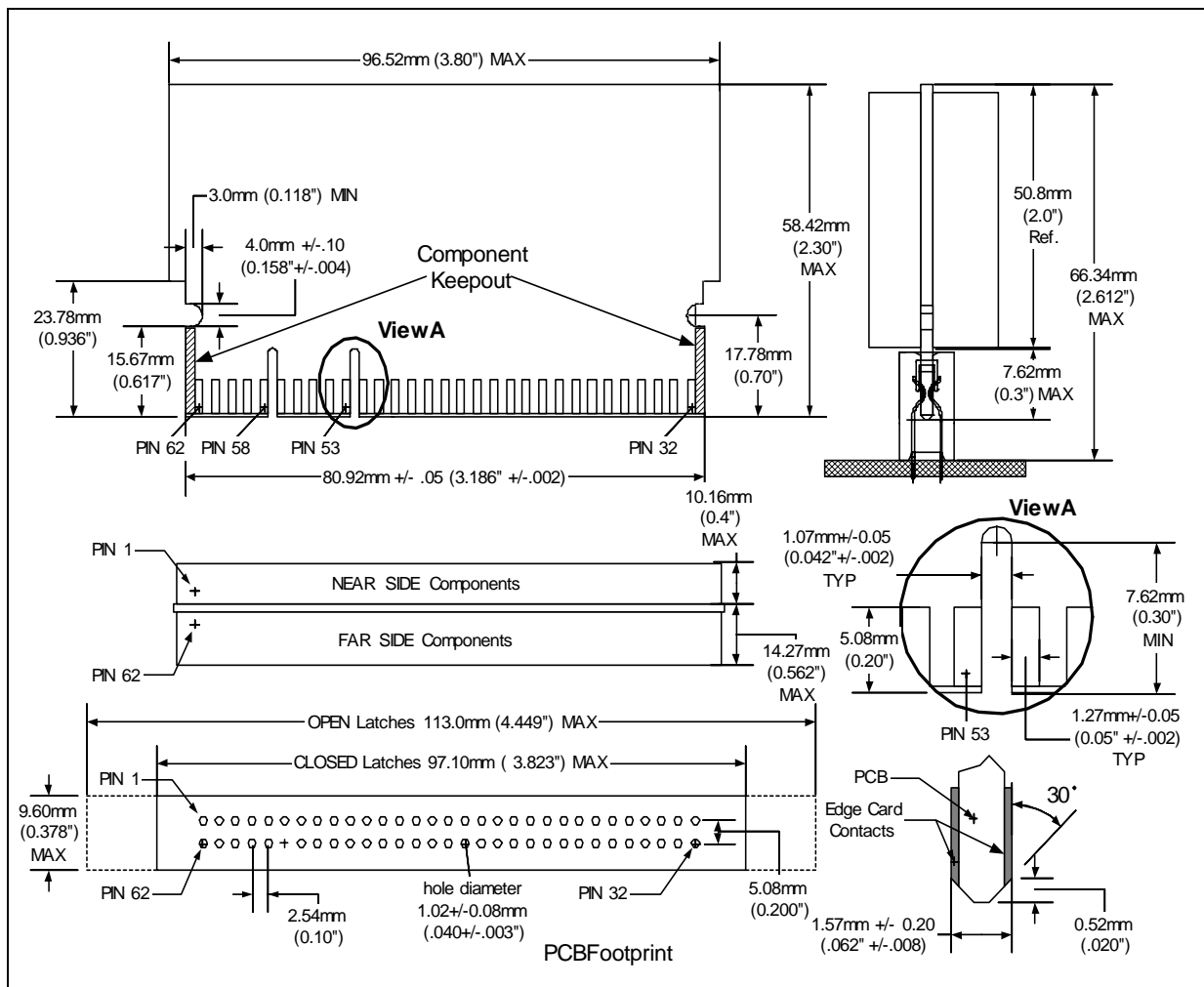
7.4 Mechanical Dimensions PROPOSED

The mechanical dimensions for the VRM 10.0 module and connector are shown in Figure 10.

7.4.1 Gold Finger Specification

The VRM board must contain gold lands (fingers) for interfacing with the VRM connector that are $1.27\text{ mm} \pm 0.05\text{ mm}$ [$0.050\text{ inches} \pm 0.002\text{ inches}$] wide by 5.08 mm [0.200 inches] minimum long and spaced $2.54\text{ mm} \pm 0.05\text{ mm}$ [$0.100\text{ inches} \pm 0.002\text{ inches}$] apart. Traces from the lands to the power plane should be a minimum of 0.89 mm [0.035 inches] wide and of a minimal length.

Figure 10. VRM 10.0 Module and Connector



8 VRM – Environmental Conditions

The VRM design, including materials, should be consistent with the manufacture of units that meet the environmental requirements specified below.

8.1 Operating Temperature **PROPOSED**

The VRM shall meet all electrical requirements when operated at the Thermal Design Current ($I_{cc}(TDC)$) over an ambient temperature of 0 °C to +60 °C with a minimum airflow of 400 LFM or 0 °C to +45 °C with a minimum airflow of 200 LFM. Operating conditions shall be considered to include 10 cycles between minimum and maximum temperature as a rate of 10 °C/hour and a dwell time of 30 minutes at extremes. Temperature and airflow measurements should be made in close proximity to the VRM.

8.2 VRM Board Temperature **REQUIRED**

To maintain the connector within its operating temperature range, the VRM board temperature, at the connector interface, shall not exceed a temperature equal to 90 °C. At no time during the operation is the VRM board permitted to exceed 90 °C within a distance of 2.54 mm [0.100 inches] from the top of the connector (equivalent to 10.16 mm [0.400 inches] from the VRM board edge). In order not to exceed the 90 °C, it is recommended that the board be constructed from 2-ounce copper cladding.

8.3 Non-Operating Temperature **PROPOSED**

The VRM shall not be damaged when exposed to temperatures between –40 °C and +70 °C. These shall be considered to include 50 cycles of minimum to maximum temperatures at 20 °C/hour with a dwell time of 20 minutes at the extremes.

8.4 Humidity **PROPOSED**

85% relative – operating

95% relative – non-operating

8.5 Altitude **PROPOSED**

3.05 km [10k feet] – operating

15.24 km [50k feet] – non-operating



8.6 Electrostatic Discharge **PROPOSED**

Testing shall be in accordance with IEC 61000-4-2.

Operating: 15 kV initialization level. The direct ESD event shall cause no out-of-regulation conditions – including overshoot, undershoot and nuisance trips of over-voltage protection, over-current protection, or remote shutdown circuitry.

Non-operating: 25 kV initialization level. The direct ESD event shall not cause damage to the VRM circuitry.

8.7 Shock and Vibration **PROPOSED**

The shock and vibration tests should be applied at the baseboard level. The VRM should not be damaged and the interconnect integrity not compromised during:

- A shock of 30 g with an 11-ms half sine wave, non-operating, to be applied in each of the orthogonal axes.
- Vibration of 0.01 g² per Hz at 5 Hz, sloping to 0.02 g² per Hz at 20 Hz and maintaining 0.02 g² per Hz from 20 Hz to 500 Hz, non-operating, applied in each of the orthogonal axes.

8.8 Electromagnetic Compatibility **PROPOSED**

Design, including materials, should be consistent with the manufacture of units that comply with the limits of FCC Class B and CISPR22 Class B for radiated emissions.

8.9 Reliability **PROPOSED**

Design, including materials, should be consistent with the manufacture of units with a Mean Time Between Failure (MTBF) of 500,000 hours of continuous operation at 55 °C, maximum outputs load, and worst case line, while meeting specified requirements. MTBF should be calculated in accordance with MIL-STD-217F or Bellcore*.

8.10 Safety **PROPOSED**

The voltage regulator is to be UL Recognized to standard UL1950 3rd Ed., including requirements of IEC950 and EN 60950. Plastic parts and printed wiring board are to be UL Recognized with 94V-0 flame class.

9 **Manufacturing Considerations**

9.1 **Lead Free (Pb Free)**

Intel recommends that designers be aware of legislation regarding the use of lead in computer products such as the European Union (EU) Restriction on Hazardous Materials directive, also known as RoHS (commonly pronounced 'rowhass'). The commission directive may be found at the following URL:

http://europa.eu.int/eur-lex/pri/en/oj/dat/2003/l_037/l_03720030213en00190023.pdf

For the latest information on RoHS please refer to the following URL:

<http://europa.eu.int/eur-lex/en>

Intel recommends that you consider Pb Free manufacturing processes and components for the module and connector.

